• (10 Points) Print your name and lab time in legible, block lettering above AND on the last page where the grading table appears.

• This exam should take up to 70 minutes to complete. You will be given at least 70 minutes, up to a maximum of 80 minutes, to work on the exam.

• This exam is closed book. Collaboration is not permitted. You may not use or access, or cause to be used or accessed, any reference in print or electronic form at any time during the exam, except three double-sided 8.5” × 11” sheets of handwritten notes having no appendage. Computing, communication, and other electronic devices (except dedicated timekeepers) must be turned off. Noncompliance with these or other instructions from the teaching staff—including, for example, commencing work prematurely or continuing beyond the announced stop time—is a serious violation of the Code of Student Conduct. Scratch paper will be provided to you; ask for more if you run out. You may not use your own scratch paper.

• The exam printout consists of pages numbered 1 through 10. When you are prompted by the teaching staff to begin work, verify that your copy of the exam is free of printing anomalies and contains all of the ten numbered pages. If you find a defect in your copy, notify the staff immediately.

• Please write neatly and legibly, because if we can’t read it, we can’t grade it.

• For each problem, limit your work to the space provided specifically for that problem. No other work will be considered in grading your exam. No exceptions.

• Unless explicitly waived by the specific wording of a problem, you must explain your responses (and reasoning) succinctly, but clearly and convincingly.

• We hope you do a fantastic job on this exam.
You may use this page for scratch work only.
Without exception, subject matter on this page will not be graded.
MT3.1 (15 Points) Consider a discrete-time LTI system $H$, as shown below:

If the input signal is characterized by $x(n) = \sum_{\ell=-\infty}^{+\infty} \delta(n - 4\ell)$ for all $n$, then the corresponding output signal is

$$y(n) = \begin{cases} 
0 & \text{if } n \text{ is even}, \\
\frac{1}{2} & \text{if } n \text{ is odd}.
\end{cases}$$

Now, suppose we apply a different input signal to the filter—an arbitrary 2-periodic signal characterized by $x(n + 2) = x(n)$ for all $n \in \mathbb{Z}$. Determine the complex-exponential discrete-time Fourier series (DTFS) expansion of the output signal $y$, which you must express in terms of $X_k$, the DTFS coefficients of the input signal $x$.

$$x(n) = \mathcal{F}\left\{ x(n) \right\} = \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} X_k e^{i k \omega_0 n} = \frac{3}{4} \sum_{k=0}^{3} \frac{e^{i k \pi n}}{2}$$

By applying the DTFS representation, we can express the output signal $y(n)$ as:

$$y(n) = Y_0 + Y_1 e^{i \frac{2\pi n}{2}} = Y_0 - \frac{1}{4} e^{i \pi n}$$

Clearly $H(0) = 1$, $H(\frac{\pi}{2}) = H(\frac{3\pi}{2}) = 0$, $H(\pi) = -1$. An arbitrary 2-periodic input $x$ has the DTFS representation $x(n) = X_0 + X_1 e^{i \pi n}$, and its corresponding output is $y(n) = X_0 H(0) + X_1 H(\pi) e^{i \pi n} = X_0 - X_1 e^{i \pi n}$. Therefore, $Y_0 = X_0$, $Y_1 = -X_1$. 

Fregs present: $0, \frac{\pi}{2}, \pi, \frac{3\pi}{2}$.
MT3.2 (40 Points) Let $\mathcal{P}_2$ denote the set of all real-coefficient polynomials of degree less than, or equal to, 2, defined on the real interval $(-1, +1)$. In other words, $\mathcal{P}_2$ consists of all zeroth-order, first-order, and second-order real-coefficient polynomials defined on the interval $(-1, +1)$:

$$\mathcal{P}_2 = \{ at^2 + bt + c \mid a, b, c \in \mathbb{R} \text{ and } t \in (-1, +1) \}.$$ 

In particular, consider the following three polynomials in $\mathcal{P}_2$:

$$\phi_0(t) = 1, \quad \phi_1(t) = \sqrt{3} t, \quad \text{and} \quad \phi_2(t) = \frac{\sqrt{5}}{2} (3t^2 - 1).$$

Let us define the inner product of two polynomials $q$ and $r$ in $\mathcal{P}_2$ as follows:

$$\langle q, r \rangle \triangleq \int_{-1}^{+1} q(t) r(t) \, dt.$$ 

(a) (18 Points) Prove the mutual orthogonality relation $\langle \phi_k, \phi_\ell \rangle = 2 \delta(k - \ell)$, for $k, \ell = 0, 1, 2$.

$$\langle \phi_0, \phi_0 \rangle = \int_{-1}^{+1} dt = 2$$

$$\langle \phi_1, \phi_1 \rangle = \int_{-1}^{+1} 3t^2 \, dt = 2 \left[ \frac{3}{2} t^3 \right]_{0}^{1} = 2 \left( \frac{3}{2} \right) = 2$$

$$\langle \phi_2, \phi_2 \rangle = \frac{\sqrt{5}}{4} \int_{-1}^{+1} (9t^4 - 6t^2 + 1) \, dt = \frac{\sqrt{5}}{4} \left. \left[ 9\frac{5}{5} t^5 - 6\frac{3}{3} t^3 + t \right] \right|_{0}^{1} = 2 \left( \frac{5}{5} - \frac{6}{3} + \frac{1}{1} \right) = 2 \left( \frac{5}{5} - \frac{6}{3} + \frac{1}{1} \right) = 2$$

$$\Rightarrow \langle \phi_2, \phi_2 \rangle = \frac{\sqrt{5}}{2} \left( \frac{9}{5} - 2 + 1 \right) = \frac{\sqrt{5}}{2} \left( \frac{2}{5} + \frac{1}{1} \right) = 2$$

Each of $\langle \phi, \phi_0 \rangle$ and $\langle \phi, \phi_2 \rangle$ is zero because each of the products $\phi(t) \phi_0(t) = \sqrt{3} t$ and $\phi(t) \phi_2(t) = \frac{\sqrt{5}}{2} (3t^2 - 1)$ is an odd function, so $\int_{-1}^{+1} \phi(t) \phi_0(t) \, dt = \int_{-1}^{+1} \phi(t) \phi_2(t) \, dt = 0$.

As for $\langle \phi_0, \phi_2 \rangle$:

$$\langle \phi_0, \phi_2 \rangle = \frac{\sqrt{5}}{2} \int_{-1}^{+1} (3t^2 - 1) \, dt = \frac{\sqrt{5}}{2} \left[ \int_{-1}^{+1} \phi_0^2(t) \, dt - \int_{-1}^{+1} \phi_2^2(t) \, dt \right] = 0$$
(b) (17 Points) Consider an arbitrary polynomial \( v \) in \( P_2 \) given by \( v(t) = at^2 + bt + c \).
We can express \( v \) as a linear combination of the mutually-orthogonal polynomials \( \phi_0, \phi_1, \) and \( \phi_2 \) as follows:
\[
v(t) = \sum_{m=0}^{2} V_m \phi_m(t) = V_0 \phi_0(t) + V_1 \phi_1(t) + V_2 \phi_2(t).
\]
Determine each of the expansion coefficients \( V_m \), where \( m = 0, 1, 2 \), in terms the polynomial coefficients \( a, b, \) and \( c \).

\[
V_m = \frac{\langle v, \phi_m \rangle}{\langle \phi_m, \phi_m \rangle} = \frac{1}{2} \left( \frac{\langle v, \phi_m \rangle}{\langle \phi_m, \phi_m \rangle} \right) \quad m = 0, 1, 2
\]

\[
V_0 = \frac{1}{2} \langle v, \phi_0 \rangle = \frac{1}{2} \int_{-1}^{1} (at^2 + bt + c) \, dt = \frac{2}{3} \left( \frac{a}{3} + c \right) \Rightarrow V_0 = \frac{a}{3} + c
\]

\[
V_1 = \frac{1}{2} \langle v, \phi_1 \rangle = \frac{\sqrt{3}}{2} \int_{-1}^{1} (at^2 + bt + c) \, dt = \frac{\sqrt{3}}{2} \left( \frac{b}{\sqrt{3}} \right) \Rightarrow V_1 = \frac{b}{\sqrt{3}}
\]

\[
V_2 = \frac{1}{2} \langle v, \phi_2 \rangle = \frac{\sqrt{5}}{4} \int_{-1}^{1} (at^2 + bt + c) \left( 3t^2 - 1 \right) \, dt = \frac{\sqrt{5}}{2} \left( \frac{3a + (3c - a)t - c}{2} \right) \int_{-1}^{1} \, dt
\]

\[
= V_2 = \frac{3a}{2\sqrt{5}} + \frac{\sqrt{5}}{6} (3c - a) - \frac{\sqrt{5}c}{2} \quad a = 3\sqrt{5}; \quad c = \frac{\sqrt{5}}{6} \quad a - \frac{\sqrt{5}c}{2} = \frac{2\sqrt{5}}{15} a
\]

(c) (5 Points) True or False? If the polynomial \( v \) in part (b) is \( \ell \)-th order, where \( \ell = 0, 1 \), then for each \( m > \ell \) the coefficient \( V_m \) must be zero. For example, if \( v \) is a first-order polynomial, then \( V_2 = 0 \). Similarly, if \( v \) is a zeroth-order polynomial (i.e., a constant), then \( V_1 = V_2 = 0 \). Provide a succinct, yet clear and convincing explanation. Depending on your approach, your answer may serve as a "sanity check" for some of the expressions you obtain in part (b).

**Zereth-Order Polynomial**: \( v(t) = c \) (\( a = b = 0 \))

\[ V_0 = \frac{a}{3} + c = c \quad V_1 = \frac{b}{\sqrt{3}} = 0 \quad V_2 = \frac{a}{\sqrt{5}} = 0 \]

**First-Order Polynomial**: \( v(t) = bt + c \) (\( a = 0 \))

\[ V_0 = \frac{a}{3} + c = c \quad V_1 = \frac{b}{\sqrt{3}} \quad V_2 = \frac{2\sqrt{5}}{15} a = 0 \]

**Intuitive Note**: \( \phi_0 \) \& \( \phi_1 \) cannot cancel any \( t^2 \) term introduced by \( \phi_2 \). Similarly, \( \phi_0 \) cannot cancel any \( t \) or \( t^3 \) term introduced by \( \phi_1 \).
MT3.3 (50 Points) This problem explores aspects of the continuous-time Fourier transform (CTFT) and the LTI processing of CT periodic signals.

(a) (10 Points) Show that the Fourier transform of the continuous-time complex exponential $e^{i\omega_0 t}$ is $2\pi \delta(\omega - \omega_0)$. That is, establish the following CTFT pair:

$$\Phi(t) = e^{i\omega_0 t} \quad \mathcal{F} \quad \Phi(\omega) = 2\pi \delta(\omega - \omega_0).$$

Hint: The CTFT analysis equation will be of no help to you. Instead, think of the frequency content of the complex exponential; from that, guess the general form of its spectrum $\Phi(\omega)$; and then plug your guessed $\Phi(\omega)$ into the CTFT synthesis equation to verify its form and determine its expression precisely.

$\Phi(t) = e^{i\omega_0 t}$ has only a single frequency: $\omega_0$. So its spectrum must be of the form $\Phi(\omega) = \alpha \delta(\omega - \omega_0)$, where $\alpha$ is a scaling constant to be determined. To determine $\alpha$, we use the CTFT synthesis equation:

$$\Phi(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} \Phi(\omega) e^{i\omega t} d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} \alpha \delta(\omega - \omega_0) e^{i\omega t} d\omega$$

By the sifting property of the Dirac delta, we know that $e^{i\omega_0 t}$ exactly $\Rightarrow \alpha = 2\pi \Rightarrow$

$$\Phi(\omega) = 2\pi \delta(\omega - \omega_0)$$

$\Phi(t) = e^{i\omega_0 t} \quad \mathcal{F} \quad \Phi(\omega)$

$\omega_0$
(b) (15 Points) Show that the Fourier transform of a uniform impulse train in time is a uniform impulse train in frequency. In particular, show that the following constitutes a CTFT pair:

\[ x(t) = \sum_{n=-\infty}^{+\infty} \delta(t - np) \quad \leftrightarrow \quad X(\omega) = \omega_0 \sum_{k=-\infty}^{+\infty} \delta(\omega - k\omega_0), \]

where \( p \) denotes the fundamental period of the time-domain impulse train, and \( \omega_0 = 2\pi/p \) represents the corresponding fundamental frequency.

Hint: Determine the continuous-time complex-exponential Fourier series representation of \( x \), and then use the result of part (a).

\[
x(t) = \sum_{k \in \mathbb{Z}} X_k e^{j k \omega_0 t} \quad \text{where} \quad \omega_0 = \frac{2\pi}{p}
\]

\[
X_k = \frac{1}{p} \int_{\langle p \rangle} x(t) e^{-j k \omega_0 t} \, dt = \frac{1}{p} \int_{-\frac{p}{2}}^{\frac{p}{2}} x(t) e^{-j k \omega_0 t} \, dt = \frac{1}{p} \quad \forall k
\]

\[
\Rightarrow x(t) = \frac{1}{p} \sum_{k \in \mathbb{Z}} e^{j k \omega_0 t} \quad \text{From part (a):} \quad e^{j \omega_0 t} \leftrightarrow 2\pi \delta(\omega - \omega_0)
\]

\[
e^{j k \omega_0 t} \leftrightarrow 2\pi \sum_{k \in \mathbb{Z}} \delta(\omega - k\omega_0)
\]

\[
\Rightarrow \sum_{k \in \mathbb{Z}} e^{j k \omega_0 t} \leftrightarrow 2\pi \sum_{k \in \mathbb{Z}} \delta(\omega - k\omega_0)
\]

\[
x(t) = \frac{1}{p} \sum_{k \in \mathbb{Z}} e^{j k \omega_0 t} \quad \Rightarrow \quad X(\omega) = \frac{2\pi}{p} \sum_{k \in \mathbb{Z}} \delta(\omega - k\omega_0) = \omega_0 \sum_{k \in \mathbb{Z}} \delta(\omega - k\omega_0)
\]
(c) (15 Points) The frequency response of a continuous-time LTI filter is shown below:

\[ H(\omega) \]

\[ B \]
\[ -2A \quad -A \quad 0 \quad A \quad 2A \quad \omega \]

Determine a reasonably simple expression for the filter’s impulse response \( h(t) \).

**Note:** You may or may not find the trigonometric identity below useful:

\[
\sin \alpha + \sin \beta = 2 \sin \left( \frac{\alpha + \beta}{2} \right) \cos \left( \frac{\alpha - \beta}{2} \right).
\]

\( H(\omega) \) can be expressed as the sum of two ideal LPF spectra:

\[
H(\omega) = \begin{array}{c}
\text{B} \\
-2A & -A & 0 & A & 2A
\end{array}
\quad + \quad \begin{array}{c}
\text{B} \\
-2A & -A & 0 & A & 2A
\end{array}
\]

\[
f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega) e^{i\omega t} d\omega = \frac{B}{2\pi} \int_{-A}^{A} e^{i\omega t} d\omega = \frac{B}{2\pi} \frac{e^{iAt} - e^{-iAt}}{it} = \frac{B}{\pi t} \sin(At)
\]

Similarly,

\[ g(t) = \frac{B}{\pi t} \sin(2At) \]

\[ H(\omega) = F(\omega) + G(\omega) \Rightarrow h(t) = f(t) + g(t) = \frac{B}{\pi t} [\sin(At) + \sin(2At)] \]

\[ \Rightarrow h(t) = \frac{2B}{\pi t} \sin\left(\frac{3At}{2}\right) \cos\left(\frac{At}{2}\right) \]
(d) (10 Points) Apply the signal $x$ in part (b) to the LTI system of part (c), where $p = \frac{8\pi}{3A}$ (so, $\omega_0 = \frac{3A}{4}$). Determine a reasonably simple expression for corresponding output $y(t)$.

$$X(\omega) \quad \uparrow \quad \uparrow \quad \uparrow \quad (\omega_0) \quad \rightarrow \quad \cdots$$

-3A/2

-3A/4

0

$\omega_0 = \frac{3A}{4}$

$2\omega_0 = \frac{3A}{2}$

-2\omega_0 = -\frac{3A}{2}

$\frac{-3A}{4} = -\omega_0$

$\omega_0 = \frac{3A}{4}$

$2\omega_0 = \frac{3A}{2}$

scales by $B$

pass by scaling $2B$

scales by $B$

Note: $e^{j\omega_t} \leftrightarrow 2\pi S(\omega)$

$\cos(\omega_0 t) \leftrightarrow \pi [S(\omega + \omega_0) + S(\omega - \omega_0)]$

$1 \leftrightarrow 2\pi S(\omega)$

You may use the blank space below for scratch work. Nothing written below this line on this page will be considered in evaluating your work.
Problem | Points | Your Score
---|---|---
Name | 10 | 10
1 | 15 | 15
2 | 40 | 40
3 | 50 | 50
**Total** | **115** | **115**

Potentially useful formulas:

**DTFS:**
\[ x(n) = \sum_{k=(p)} X_k e^{i\omega_0 n} \quad \longleftrightarrow \quad X_k = \frac{1}{p} \sum_{n=(p)} x(n) e^{-i\omega_0 n}. \]

**CTFS:**
\[ x(t) = \sum_{k=-\infty}^{+\infty} X_k e^{i\omega_0 t} \quad \longleftrightarrow \quad X_k = \frac{1}{p} \int_{(p)} x(t) e^{-i\omega_0 t} dt. \]

**DTFT:**
\[ x(n) = \frac{1}{2\pi} \int_{(2\pi)} X(\omega) e^{i\omega n} d\omega \quad \longleftrightarrow \quad X(\omega) = \sum_{n=-\infty}^{+\infty} x(n) e^{-i\omega n}. \]

**CIF1:**
\[ x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(\omega) e^{i\omega t} d\omega \quad \longleftrightarrow \quad X(\omega) = \int_{-\infty}^{+\infty} x(t) e^{-i\omega t} dt. \]