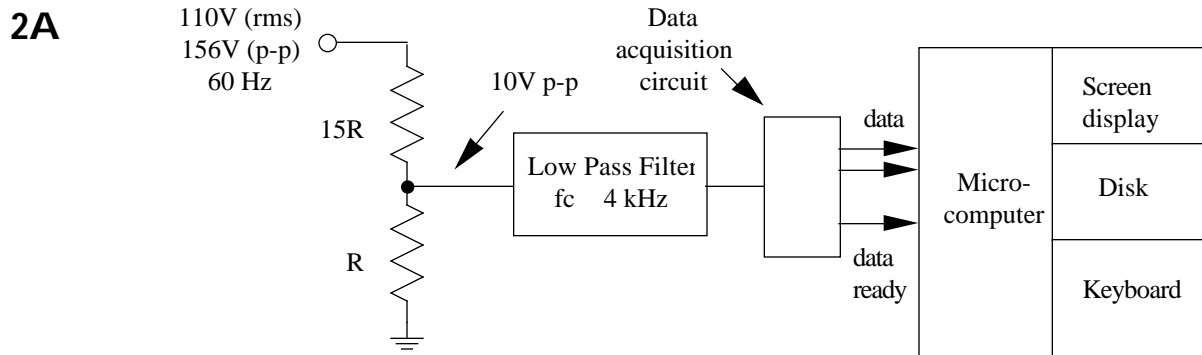


UNIVERSITY OF CALIFORNIA
 College of Engineering
 Electrical Engineering and Computer Sciences Department
 145M Microcomputer Interfacing Lab
 Final Exam Solutions May 18, 1992

- 1A Nyquist Theorem:** To recover a waveform from its sampled values, the highest frequency present must \leq one-half the sampling frequency.
- 1B Discrete Fourier Transform:** Transform for determining the amplitude of the frequency components of a periodically sampled waveform.
- 1C Differential Linearity Error (of an A/D converter):** Difference between the spacing of neighboring transition voltages and their average spacing. [4 points off if step size or transition voltage not mentioned] [4 points off if absolute or relative accuracy was defined]
- 1D Anti-Aliasing Filter:** Low-pass filter used to block frequencies above one-half the sampling frequency and thereby prevent aliasing.
- 1E Power Amplifier:** Amplifier having high power or current output, and required to drive an actuator such as a speaker or heater. [2 points off if high power or current output not mentioned]
- 1F Digital Filter:** Filter whose output is a linear combination of previous input and output values.



[3 points off if 156 V p-p sent directly into acquisition circuit]
 [3 points off if low-pass filter omitted]

- 2B** $f = 0.01 \text{ Hz}$, $S = 1/f = 100 \text{ sec}$, $N = 100 \text{ sec} \times 10 \text{ kHz} = 10^6 \text{ samples}$
 ($S = 50 \text{ sec}$, $N = 5 \times 10^5 \text{ samples}$ also acceptable)
- 2C** F_0 corresponds to 0 Hz or dc.
- 2D** 60 Hz corresponds to F_{6000} and F_{N-6000} where $N = 10^6$
- 2E** Since the distortion has a period of 60 Hz, only multiples of 60 Hz will be present.
 The highest harmonic n_{max} that can pass the anti-aliasing filter is $5000 \text{ Hz} / 60 \text{ Hz} = 80$.
 The n th harmonic will be at F_{6000n} and $F_{N-6000n}$
 [3 points off if only frequencies given]
 [6 points off if only F_{6000} and F_{N-6000} given]
 [6 points off if answer says that all Fourier amplitudes are non-zero]
- 2F**
- 1) Data acquisition circuit samples waveform, digitizes, and sets data ready line
 - 2) When program detects data ready line, it reads data, stores data, and resets data ready line
 - 3) Loop back to step 1 until 10^6 values taken
 - 4) Multiply values by Hanning window
 - 5) Compute the FFT
- [2 points off for each step missing]

3A The first harmonic could be anywhere in the 59.9 to 60.1 Hz range, which corresponds to 20 potentially non-zero Fourier coefficients (40 filters) from F_{5990} to F_{6010} .
 The range of the 80th harmonic will involve $40 \times 80 = 3200$ filters.
 The average number of filters per harmonic = 1600.
 The total number of filters is then $1600 \times 80 = 128,000$ (much less than the total number of 500,000 real plus 500,000 imaginary components)
 [4 points off for an answer of 2-5 filters]
 [3 points off for answers of 10-40 filters or 10^6 filters]
 [2 points off for an answer from 3000 to 7000 filters]
 [2 points off if the 59.9 to 60.1 Hz range was neglected]

3B The real FIR filter is given by $F_n = \sum_{k=0}^{N-1} f_k \cos(2\pi nk/N)$ where $n = 100f$ and $N = 10^6$.
 [2 points off if frequency f missing]

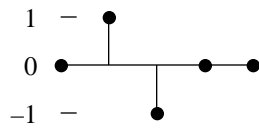
The real IIR filter is given by $F_n(t + t) = [F_n(t) - f_0 + f_N] \cos(2\pi n/N)$ where $n = 100f$ and $N = 10^6$.

3C

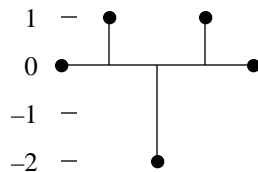
- 1 Take 10^6 samples (f_0 to f_{N-1}) and store in memory
- 2 Compute all needed coefficients in a loop
- 3 Start conversion and read new sample when ready
- 4 Delete oldest value (f_0), store new value (f_N) after last value (f_{N-1}), and shift all f_k to f_{k-1}
- 5 Loop back to step 2

[3 points off for each missing step]

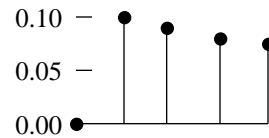
4A
 $y_0 = 0$ $y_1 = 1$ $y_2 = -1$
 $y_3 = 0$ $y_4 = 0$



4B
 $y_0 = 0$ $y_1 = 1$ $y_2 = -2$
 $y_3 = 1$ $y_4 = 0$



4C
 $y_0 = 0$ $y_1 = 0.100$ $y_2 = 0.091$
 $y_3 = 0.081$ $y_4 = 0.073$

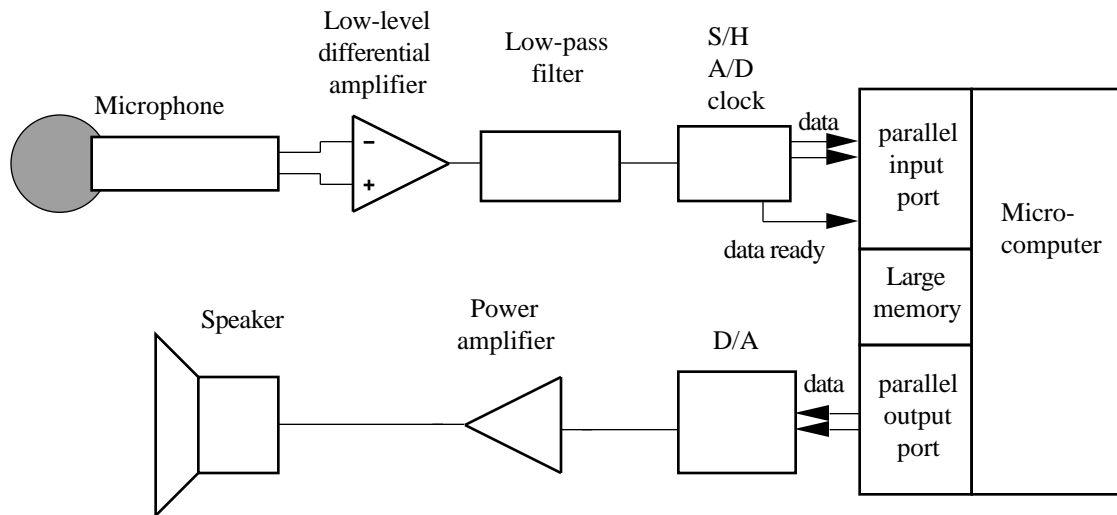


4D Second derivative — b a — FIR
 First derivative — a b — FIR
 Low pass filter — c c — IIR

5A

- 1 With an input impulse, sample the output waveform $h(t)$ and perform the FFT.
- 2 Perform the FFT of the desired output $y(t)$
- 3 Compute the required input $u(t)$ using $u(t) = \text{FFT}^{-1} \frac{\text{FFT}(y)}{\text{FFT}(h)}$

5B



[1 point off for each amplifier missing] [1 point off for missing low pass filter]

5C

The digital filter is given by

$$y_i = (1 - \alpha)x_{i-1} + \alpha y_{i-1} \quad \alpha = \exp(-T/RC)$$

where T is the cycle time (sampling x_{i-1} + computing y_i)

Gain vs. frequency is similar to the analog filter.

Phase is similar but shifted later by T.

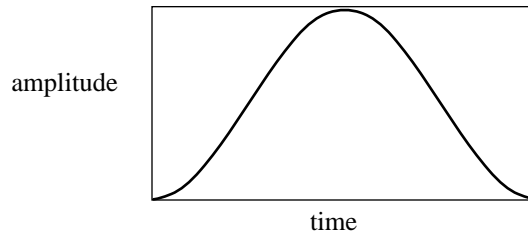
[1 point off for $\exp(-1/RC)$] [2 points off if processing delay omitted]

[3 points off if not related to RC]

5D

The simple rectangular window allows a discontinuity between the ends of the sampled values.

To avoid this, need to multiply the time samples by a function that smoothly approaches zero at both ends:



145M Numerical Grades:

	6/9 x Lab	Lab Partic.	Midterm	Final	Total
Average	564	100	78	175	916 (B+)
rms	24	0	17	20	57
Maximum	600	100	100	200	1000

145M Letter Grade Distribution

Letter Grade	Course Totals (1000 max)
A+	980*, 969
A	968, 965, 960, 957, 949*
A-	none
B+	929, 919, 913, 904, 903
B	891, 884
B-	none
C	840
D	753

* Graduate students