`Solutions for Midterm #2 - EECS 145M Spring 2008

1.1 Frequency aliasing is caused by sampling at a frequency f_s that is less than twice the highest frequency f_{max} in the waveform.

It may be avoided by using an analog low-pass filter to eliminate frequencies above $f_s/2$.

[4 points off for increasing the sampling frequency, since f_{max} is not known]

1.2 Spectral leakage is caused when frequency components are not sampled for a whole number of cycles, which results in a discontinuity between the last sample and the "next sample", which is also the first sample.

It is avoided by multiplying the sampled values by a windowing function that has zero value and zero slope at the ends of the sampling window.

[3 points off for a solution that samples for a longer time- this does not eliminate the discontinuity and long-range spectral leakage will still occur]

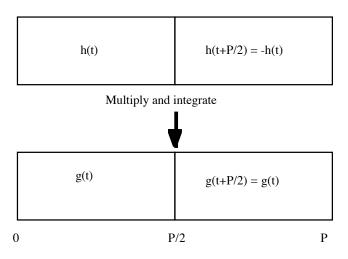
[3 points off if the only answer is to sample for an integer number of cycles- nothing is known about the frequency components present]

2.1 H(f) can only be non-zero at multiples of the repeat frequency = n/P

[4 points off for describing or sketching only the first harmonic]

2.2 All the components in the Fourier integral have periods that are integer multiples of P, so integrating from 0 to P is all that is necessary.

$$H(f) = \int h(t)g(t)dt \quad g(t) = \exp(-j2\pi nt / P)$$



 $g(t) = \exp(-j2\pi nt/P)$ For *n* even, $g(t+P/2) = \exp(-j2\pi n[t+P/2]/P) = \exp(-j2\pi nt/P-j2\pi n/2) = \exp(-j2\pi nt/P) = g(t)$

For even *n*, the harmonic factor $\exp(-j2\pi nt/P)$ is the same in the interval from t = 0 to P/2 as it is in the interval from t = P/2 to *P*. However, h(t) has the opposite sign from t = 0 to P/2 as it does from t = P/2 to *P* and the two inner products cancel.

[Full credit for showing that the integral from 0 to P could be split into two integrals that cancel][4 exams out of 17 had part **2.2** correct]

[5 points off for a valiant but incomplete attempt]

[the most common approach was to integrate both h(t)g(t) and h(t+P/2)g(t+P/2) from 0 to *P*, without following up with a convincing argument that each was zero for even *n*.]

3.1 Need to show Butterworth filter, S/H amplifier, A/D, timer, computer [2 points off for each missing component] $G_1 = 0.99 f_1/f_c = 0.823, f_c = 10 \text{ kHz}/0.823 = 12.15 \text{ kHz}$ **3.2** $G_2 = 0.001 f_2/f_c = 1.995; f_2 = 24.24 \text{ kHz}$ $f_s > f_1 + f_2 = 34.24 \text{ kHz}$ [3 points off for $f_s > 2f_c = 24$ kHz. This is not sufficient] [5 points off for $fs > 2f_1 = 20$ kHz. This assumes a perfect rejection above 10 kHz. **3.3** T = $1/(f * 2^{17} * \pi) = 1/(10 \text{ kHz} * 131,072 * 3.1416) = 0.24 \text{ ns}$ [2 points off for a minor numerical error] [5 points off for not solving for T] **3.4** For S = 0.4 s, $\Delta f = 2.5$ Hz. [S = 0.3 to 0.5 seconds was accepted for full credit][5 points off for S = 0.1 seconds] [3 points off for S = 0.2 seconds] **3.5** $34.24 \text{ kHz} \ge 0.4 \text{ sec} = 13,696 \text{ (next power of 2 is 16,384).}$ 3.6 Set counter to sample A/D for 0.4 sec at 34.24 kHz

Start sampling output of Butterworth filter

When done, multiply samples by Hann window

Take FFT

[2 points off for no Hann or similar window]

[2 points off for not giving the frequency that corresponds to H_n]

Each Fourier frequency component H_n corresponds to 2.5xn Hz.

| Problem | max | average | rms |
|---------|-----|---------|-----|
| 1 | 20 | 19.5 | 1.4 |
| 2 | 20 | 13.5 | 3.2 |
| 3 | 60 | 51.5 | 6.4 |
| total | 100 | 84.6 | 8.3 |

Grade distribution:

| Range | number | approximate |
|-------|--------|--------------|
| | | letter grade |
| 65-69 | 1 | C+ |
| 70-74 | 1 | B- |
| 75-79 | 3 | В |
| 80-84 | 1 | B+ |
| 85-89 | 8 | A- |
| 90-94 | 1 | А |
| 95-99 | 2 | A+ |
| 100 | 0 | A+ |