

EECS 120  
Final Exam  
Fri. May 16, 2014  
0810 - 1100 am

Name: \_\_\_\_\_

SID: \_\_\_\_\_

- Closed book. Three single sided 8.5x11 inch pages of formula sheet. No calculators.
- There are 8 problems worth 200 points total. There may be more time efficient methods to solve problems.

Problem	Points	Score
1	24	
2	29	
3	14	
4	29	
5	24	
6	34	
7	24	
8	22	
TOTAL	200	

In the real world, unethical actions by engineers can cost money, careers, and lives. The penalty for unethical actions on this exam will be a grade of zero and a letter will be written for your file and to the Office of Student Conduct.

Tables for reference:

$\tan^{-1} 0.1 = 5.7^\circ$	$\tan^{-1} 0.2 = 11.3^\circ$
$\tan^{-1} \frac{1}{2} = 26.6^\circ$	$\tan^{-1} 1 = 45^\circ$
$\tan^{-1} \frac{1}{3} = 18.4^\circ$	$\tan^{-1} \frac{1}{4} = 14^\circ$
$\tan^{-1} \sqrt{3} = 60^\circ$	$\tan^{-1} \frac{1}{\sqrt{3}} = 30^\circ$
$\sin 30^\circ = \frac{1}{2}$	$\cos 60^\circ = \frac{\sqrt{3}}{2}$
$\cos 45^\circ = \frac{\sqrt{2}}{2}$	$\sin 45^\circ = \frac{\sqrt{2}}{2}$

$20 \log_{10} 1 = 0dB$	$20 \log_{10} 2 = 6dB$	
$20 \log_{10} \sqrt{2} = 3dB$	$20 \log_{10} \frac{1}{2} = -6dB$	
$20 \log_{10} 5 = 20dB - 6dB = 14dB$	$20 \log_{10} \sqrt{10} = 10 dB$	
$1/e \approx 0.37$	$1/e^2 \approx 0.14$	$\sqrt{2} \approx 1.41$
$1/e^3 \approx 0.05$	$\sqrt{10} \approx 3.16$	$\sqrt{3} \approx 1.73$

**Problem 1 LTI Properties (24 pts)**

[24 pts] Classify the following systems, with input  $x(t)$  (or  $x[n]$ ) and output  $y(t)$  (or  $y[n]$ ). In each column, write “yes”, “no”, or “?” if the property is not decidable with the given information. (+1 for correct, 0 for blank, -0.5 for incorrect).

System	Causal	Linear	Time-invariant	BIBO stable
a. $y(t) = x(t) * \sum_{n=0}^{\infty} \delta(t - 4n)$				
b. $y(t) = x(t) \cdot \sum_{n=-\infty}^{\infty} \delta(t - 4n)$				
c. $y[n] = x[n] * 0.9^n u[n]$				
d. $y[n] = \sum_{m=0}^{\infty} x[m] 0.9^m$				
d. $y(t) = x(t) * [\delta(t + 1) + e^{-2t} u(t)]$				
e. $y(t) = x(t) * [\frac{d}{dt} \delta(t - 1) + e^{-2t} u(t)]$				

**Problem 2 Short Answers (29 pts)**

Answer each part independently. Note  $\Pi(t) = u(t + \frac{1}{2}) - u(t - \frac{1}{2})$ .

[3 pts] a. Evaluate  $\delta(t + \frac{1}{4}) * \cos(2\pi t)u(t) =$  \_\_\_\_\_

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[4 pts] b. Sketch  $\Pi(t - 1) * \Pi(2t)$

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[4 pts] c. Given an LTI system with input  $x(t) = \Pi(t - 4.5)$  and output  $y(t) = \Pi(2t) - \Pi(2(t - 1))$ , find the impulse response of the system,

$h(t) =$  \_\_\_\_\_. (Hint: sketch  $x(t)$  and  $y(t)$ )

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[4 pts] d. Given  $x(t) = \sum_{n=-\infty}^{\infty} [\delta(t - 2n) + \frac{1}{2}\delta(t - 2n + 1)]$ , find  $X(j\omega)$  the Fourier Transform of  $x(t)$ .

$X(j\omega) =$  \_\_\_\_\_

[4 pts] e. What is the energy in the time signal  $\frac{\sin(100\pi t)}{t}$ ? \_\_\_\_\_

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[6 pts] f. A system with input  $x(t)$  and output  $y(t)$  is described by the following differential equation:  
 $\frac{d^2}{dt^2}y + 2\frac{d}{dt}y + y = \frac{d}{dt}x + 2x.$

Assuming zero initial conditions, find the impulse response for this system.

$h(t) =$  \_\_\_\_\_

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[4 pts] g. Given the bilateral Laplace transform  $X(s) = \frac{1}{s-2}$  and the region of convergence is to the left of the pole at  $s = 2$ , find the inverse Laplace transform,

$x(t) =$  \_\_\_\_\_

**Problem 3 Laplace Transform (14 pts)**

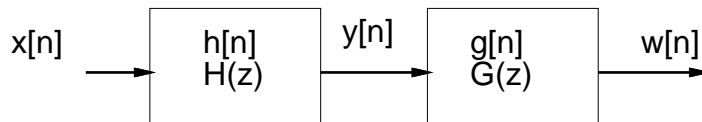
[14 pts] An LTI system has input  $x(t)$ , output  $y(t)$  and transfer function

$$H(s) = \frac{1}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

For input  $x(t) = (1 + \cos \omega_n t)u(t)$ , find the steady-state solution  $y(t)$  for large  $t$ .

$y(t) =$  \_\_\_\_\_

**Problem 4. Z transform (29 pts)**



[10 pts] a. Consider an LTI causal system with impulse response  $h[n] = (2 - (\frac{1}{2})^n)u[n]$ . Find  $g[n]$  such that  $h[n] * g[n] = \delta[n]$ .

$g[n] =$  \_\_\_\_\_

[3 pts] b. Show by direct calculation that the  $g[n]$  from part a. above is the inverse of  $h[n]$  from part a.

[4 pts] c. Show that  $g[n]$  is the impulse response of a stable system.

[12pts] d. Consider an LTI causal system with Z transform

$$H(z) = \frac{z(z - 2)}{z - 3/4}$$

Find a **stable**  $G(z)$  such that  $|H(e^{j\Omega})G(e^{j\Omega})| = 1$  for all  $\Omega$ .

$G(z) =$  \_\_\_\_\_

**Problem 5. Z Transform (24 pts)**

A causal system with input  $x[n]$  and output  $y[n]$  is described by the difference equation:

$$y[n] + 0.3y[n - 1] - 0.4y[n - 2] = x[n] - x[n - 1]$$

[12 pts] a. Find  $Y(z)$  and  $y[n]$  for  $x[n] = 0$  (ZIR), with  $y[-2] = 4$  and  $y[-1] = 2$ .

$$Y(z) = \underline{\hspace{2cm}} \qquad y[n] = \underline{\hspace{2cm}}$$

[12 pts] b. Find  $Y(z)$  and  $y[n]$  for  $x[n] = u[n]$  (ZSR).  $y[-2] = 0$  and  $y[-1] = 0$ .

$$Y(z) = \underline{\hspace{2cm}} \qquad y[n] = \underline{\hspace{2cm}}$$

**Problem 6. Digital Filter (34 pts)**

A continuous time causal LTI filter has transfer function

$$H(s) = 4 \frac{s + 1}{s + 4}$$

[4 pts] a. Find the linear differential equation with constant coefficients with input  $x(t)$  and output  $y(t)$  which has transfer function  $H(s)$  (assume zero initial conditions).

LDE: \_\_\_\_\_

[4 pts] b. Using the backward difference approximation for the derivative (i.e.

$$\frac{dy}{dt} \approx \frac{y[n] - y[n-1]}{T},$$

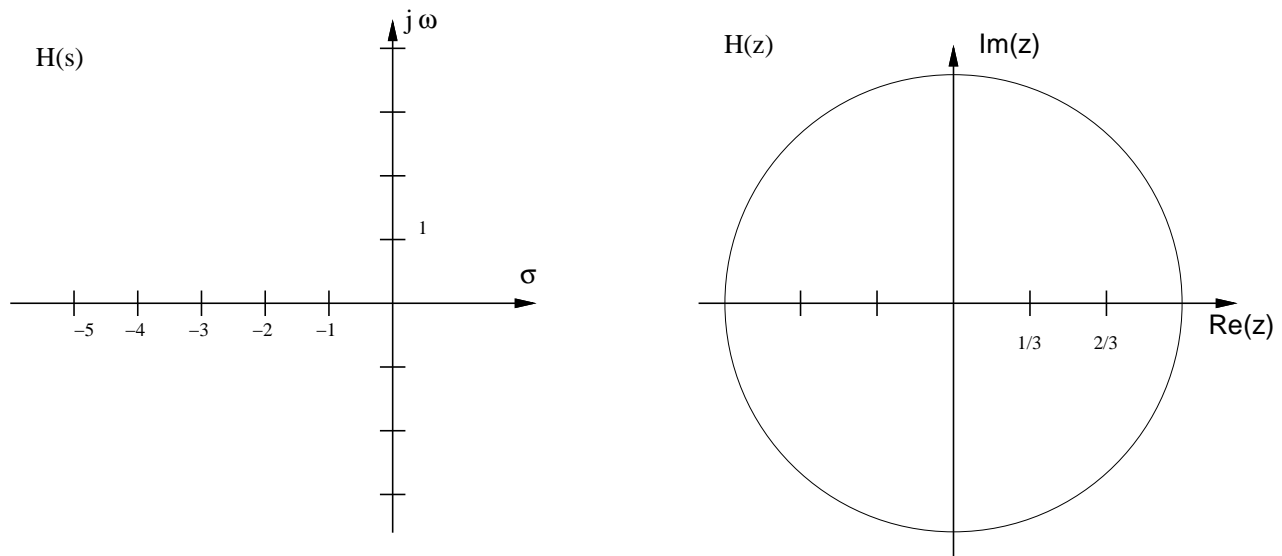
with  $T = \frac{1}{2}$ , find the linear difference equation approximation with input  $x[n]$  and output  $y[n]$ .

LDE: \_\_\_\_\_

[4 pts] c. Assuming zero initial conditions, find the Z transform for the LDE in part b.

$H(z) =$  \_\_\_\_\_

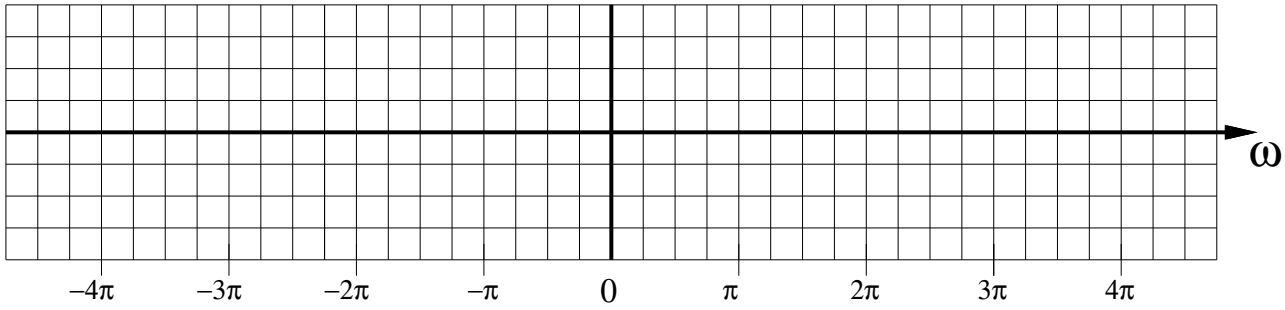
[4 pts] d. Draw pole-zero diagrams for  $H(s)$  in the s-plane and  $H(z)$  in the z-plane.





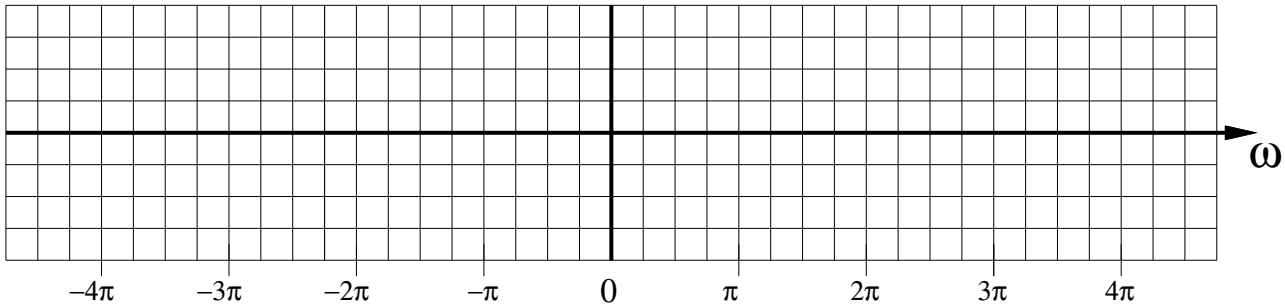
[8 pts] e. Sketch the magnitude of frequency response of the continuous time system, labelling key amplitudes.

$|H(j\omega)|$ :



[8 pts] f. Sketch the magnitude of frequency response of the discrete time system, noting that  $T = \frac{1}{2}$  sec., labelling key amplitudes.

$|H(e^{j\omega T})|$ :

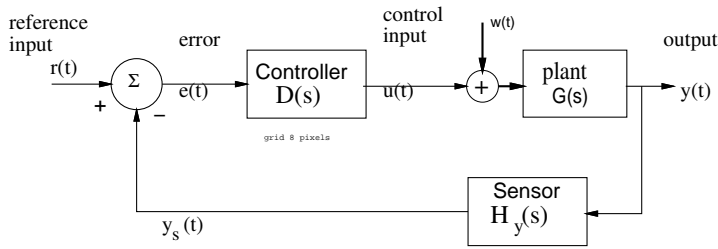


[2 pts] g. Briefly explain the reasons for differences between the magnitudes of the CT and DT frequency responses.

Some possibly useful constants:

$\pi \approx 3.14$	$2\pi \approx 6.3$
$3\pi \approx 9.42$	$4\pi \approx 12.6$
$\sqrt{2} \approx 1.4$	$\sqrt{3} \approx 1.7$
$\sqrt{5} \approx 2.2$	$\sqrt{10} \approx 3.2$
$\sqrt{8} \approx 2.8$	$\sqrt{17} \approx 4.1$
$\sqrt{20} \approx 4.5$	$\sqrt{26} \approx 5.1$

**Problem 7. Control (24 pts)**



[3 pts] a. Find the transfer function  $\frac{E(s)}{R(s)}$  in terms of  $D, G, H_y$ .

$$\frac{E(s)}{R(s)} = \underline{\hspace{2cm}}$$

[3 pts] b. Find the transfer function  $\frac{E(s)}{W(s)}$  in terms of  $D, G, H_y$ .

$$\frac{E(s)}{W(s)} = \underline{\hspace{2cm}}$$

For the system above, let  $D(s) = k_p$ ,  $H_y(s) = \frac{s+1}{s}$ , and  $G(s) = \frac{1}{s^2+as+b}$ .

[10 pts] c. With  $r(t) = 0$ , determine trend of  $e(t)$  as  $t \rightarrow \infty$  with respect to a step disturbance input  $w(t)$ .

$$e(t) \rightarrow \underline{\hspace{2cm}}$$

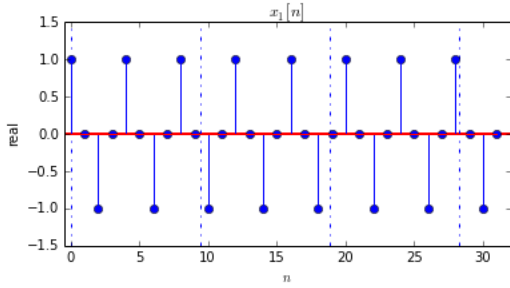
[8 pts] d. With  $w(t) = 0$ ,  $H_y(s) = 1$ ,  $D(s)G(s) = \frac{200}{(s+1)^2(s+10)^2}$ , determine  $\omega_c$  for which  $|D(j\omega_c)G(j\omega_c)| \approx 1$  and the approximate phase margin. (Hint for small angles  $\tan^{-1} 0.1 \approx 0.1 \text{ rad} \approx 5.7^\circ$ .)

$$\omega_c = \underline{\hspace{2cm}}$$

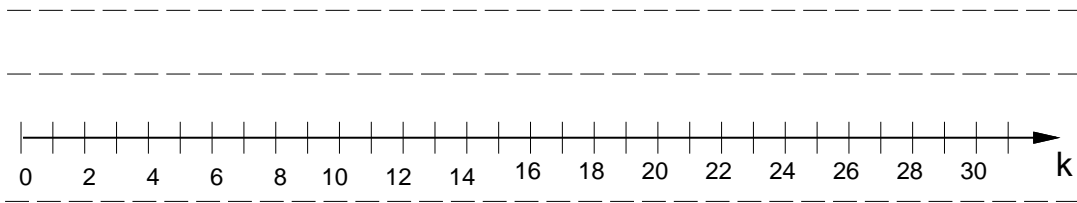
phase margin (specify rad or degrees)  $\underline{\hspace{2cm}}$

**Problem 8. DFT problem or pole-zero match (22 pts)**

[11 pts] a. Given  $x_1[n] = \cos(2\pi\frac{n}{4})$  as shown:

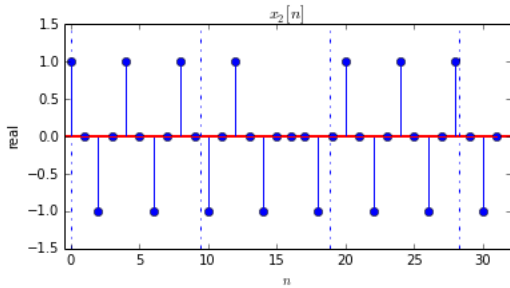


sketch  $X_1[k]$ , the 32 point DFT of  $x_1[n]$ , labelling amplitudes.

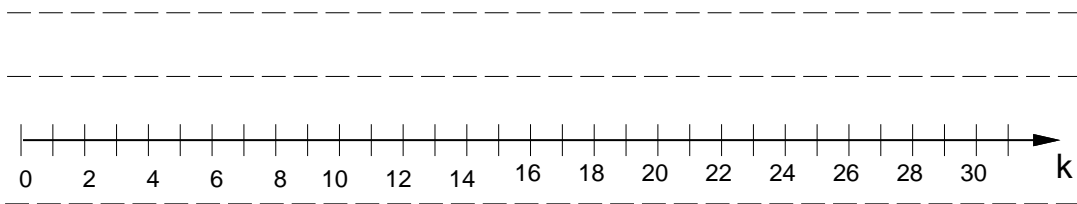


$X_1[k]$ : -----

[11 pts] b. Given  $x_2[n] = \cos(2\pi\frac{n}{4}) - \delta[n - 16]$  as shown:



sketch  $X_2[k]$ , the 32 point DFT of  $x_2[n]$ , labelling amplitudes.



$X_2[k]$ : -----

Area for scratch work.