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Spring 2000

**EECS 121 — MIDTERM**  
(7:00-9:00 p.m., 8 Wednesday 2000)

**Please explain your answers carefully. There are 100 total points and Question 4, part c) is a bonus.**

**Problem 1 (30 points)**

[8 pts.] **1a)** Argue that for any binary code satisfying the prefix-free condition, the codeword lengths  $\{l_i\}$  must satisfy the Kraft's inequality:

$$\sum_i 2^{-l_i} \leq 1.$$

[6 pts.] **b)** Is it true that for *any* source, the codeword lengths for the binary Huffman code must satisfy Kraft's inequality with equality? Explain.

[6 pts.] **c)** Suppose now the coded symbols are from a general alphabet of size  $D$ . The Kraft inequality becomes:

$$\sum_i D^{-l_i} \leq 1.$$

Is it true that the Huffman code must satisfy Kraft's inequality with equality? Explain.

[10 pts.] **d)** Consider a source for which the letter probabilities are of the form  $2^{-k}$ , where  $k$  is an integer. Construct the Huffman code and give the corresponding codeword lengths. Justify that the code is optimal.

**Problem 2 (30 points)**

Let  $\{X(t)\}$  be a zero-mean WSS Gaussian process with autocorrelation function  $R_x(\tau) = e^{-|\tau|}$ .

**[6 pts.] a)** Find its power spectral density.

**[8 pts.] b)** Suppose we sample this process every  $T$  seconds. Is the resulting discrete-time process Gaussian? WSS? If so, compute its autocorrelation function.

**[8 pts.] c)** Let  $\{Y_n\}$  be the sampled process. We perform DPCM quantization by LLSE prediction of  $Y_n$  from  $Y_{n-1}$ . Find the distribution of the residual error  $Y_n - \hat{Y}_n$ .

**[8 pts.] d)** The residual error is quantized by a single bit quantizer to values  $\pm\Delta$ . Find the optimal choice of  $\Delta$  as a function of  $T$ . What happens when  $T \rightarrow 0$ ?

**Problem 3 (20 points)**

Here is one way to simulate white noise. Let  $\{W_n\}$  be iid. rv's with  $P(W_n = 1) = P(W_n = -1) = \frac{1}{2}$ . For each  $K$ , define the continuous process  $\{W^{(k)}(t)\}$  for  $t \geq 0$  as follows:

$$W^{(k)}(t) = W_n \sqrt{K} \quad \text{for } \frac{n}{K} \leq t \leq \frac{n+1}{K}, \quad n = 0, 1, 2, \dots$$

For large  $K$ , this can be used to approximate  $\{W(t)\}$ .

[4 pts.] a) Sketch a typical sample path of  $\{W^{(k)}(t)\}$ .

[10 pts.] b) Compute  $\text{Var}[W^{(k)}(t)]$  and  $\text{Var}\left[\int_0^1 W^{(k)}(t) dt\right]$ . Based on this calculation, explain why while

$$\text{Var}[W(t)] = \infty, \quad \text{Var}\left[\int_0^1 W(t) dt\right] \text{ is finite.}$$

[6 pts.] c) A student does not like the fact that  $\text{Var}[W(t)] = \infty$ . He decides to use instead the approximation

$$Y^{(k)}(t) = W_n \quad \text{for } \frac{n}{K} \leq t \leq \frac{n+1}{K}, \quad n = 0, 1, 2$$

What is wrong with this noise model for  $K$  large?

**Problem 4 (20 points)**

A data stream is partitioned into blocks of two bits. A block is modulated onto a signal waveform, say on  $[0, 1]$ . Consider two modulation schemes:

- 1) Scheme A: The two bits are modulated into a 4-level PAM, with the 4 levels equally spaced and symmetric about 0.
- 2) Scheme B: The signal waveform is composed by separately modulating each bit into a 2-level PAM. The waveform for the first bit is on  $\left[0, \frac{1}{2}\right]$ , and the one for the second bit on  $\left[\frac{1}{2}, 1\right]$ . The levels are symmetric about 0.

**[6 pts.] a)** Sketch the possible signal waveforms for both schemes.

**[14 pts.] b)** Find an orthonormal basis for each scheme (on  $[0, 1]$ ). What are the dimensions of the signal space? Sketch a geometric representation of the signal constellation.

**[6 pts.] c)** BONUS: An important measure of a modulation scheme is the minimum distance between the constellation points. For a given minimum distance  $d$ , find the average energies required in both schemes. Which scheme is better in this respect? (You can assume that each of the four possible messages is equally likely.)