

UNIVERSITY OF CALIFORNIA, BERKELEY
College of Engineering
Department of Electrical Engineering and Computer Sciences

EE 105: Microelectronic Devices and Circuits

Fall 2009

MIDTERM EXAMINATION #1

Time allotted: 45 minutes

NAME: _____

STUDENT ID#: _____

INSTRUCTIONS:

- 1. SHOW YOUR WORK. (Make your methods clear to the grader!)**
Specially, while using chart, make sure that you indicate how you have got your numbers. For example, if reading off mobility, clearly write down what doping density that corresponds to.
- 2. Clearly mark (underline or box) your answers.**
- 3. Specify the units on answers whenever appropriate.**

SCORE: 1 _____ / 15

2 _____ / 15

3 _____ / 20

Total _____ / 50

PHYSICAL CONSTANTS

Description	Symbol	Value
Electronic charge	q	1.6×10^{-19} C
Boltzmann's constant	k	8.62×10^{-5} eV/K
Thermal voltage at 300K	$V_T = kT/q$	0.026 V

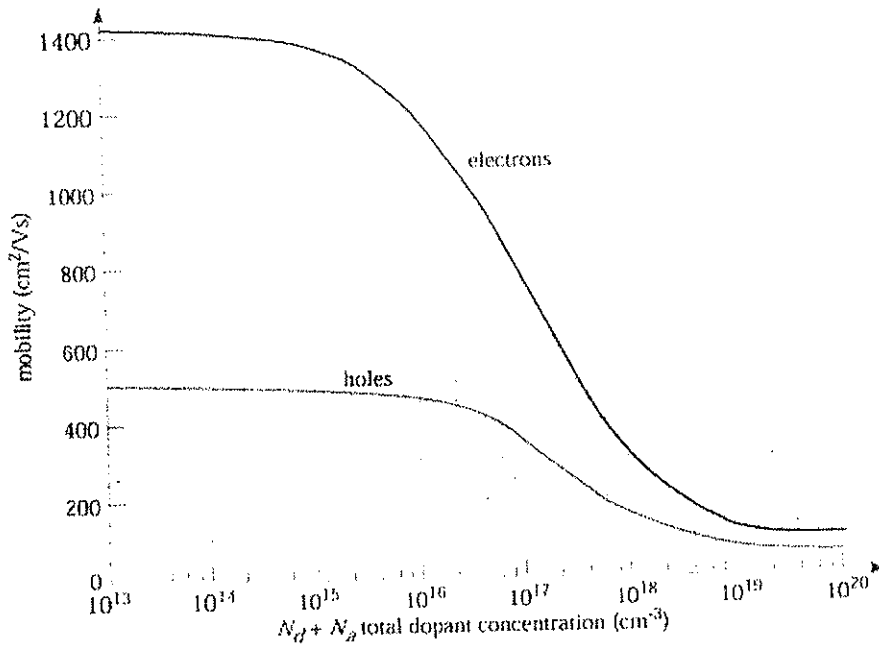
PROPERTIES OF SILICON AT 300K

Description	Symbol	Value
Band gap energy	E_G	1.12 eV
Intrinsic carrier concentration	n_i	10^{10} cm ⁻³
Dielectric permittivity	ϵ_{Si}	1.0×10^{-12} F/cm

USEFUL NUMBERS

$V_T \ln(10) = 0.060$ V at $T=300$ K
 $\exp(30) \sim 10^{13}$

Electron and Hole Mobilities in Silicon at 300K



Prob 1. [15 pts] Let us assume that we have a Si sample of length $16 \mu\text{m}$ and a cross sectional area of $10 \mu\text{m}^2$ at room temperature. Also assume that we apply 1V across this sample.

(a) If it is desired to have an electron density that is 10^5 times the intrinsic density, [4 pt]

- (i) What kind of dopant atoms is needed? Can you give an example?
- (ii) What doping density will be required?

(i) Donors \rightarrow group V \rightarrow Phosphorous

(ii) $n = 10^{15} / \text{cm}^3 \approx N_D$

(b) Estimate the drift velocity of this sample. [5 pt]

$$v = \mu_n E = 1350 \times \frac{1(\text{V})}{16 \times 10^{-4} (\text{cm})}$$

\uparrow

from chart corresponding to 10^{15}

$$\approx 8.4 \times 10^5 \text{ cm/sec}$$

- (c) Now let us assume that we counter-dope the sample with opposite type of dopants with a density that is 3 times more than the previous dopant density. Estimate the resistance and current flowing in this sample under these conditions. [6 pt]

material becomes p-type: $p = (3 \cdot 1) \times 10^{15} = 3 \times 10^{15} \text{ cm}^{-3}$

$$R = \rho \frac{L}{A} = \frac{L}{\sigma A}$$

$\sigma = e \mu_p p \rightarrow \mu_p \sim 500$ corresponding to $N_A + N_D = 9 \times 10^{15} \text{ cm}^{-3}$

$$R = \frac{1}{1.6 \times 10^{19} \times 500 \times 3 \times 10^{15}} \times \frac{16 \times 10^{-4}}{20 \times (10^{-4})^2}$$

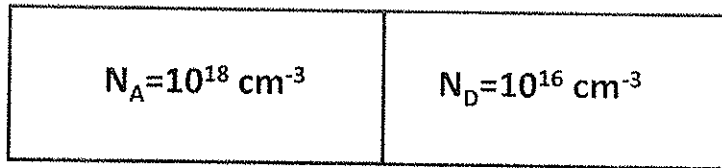
$$= \frac{1}{10^{19} \times 10^{18} \times 10^{-4}}$$

$$R = 10^3 \Omega$$

$$I = \frac{10}{10^3 \Omega} = 10^{-2} \text{ A}$$

$$I = 10 \text{ mA}$$

Prob 2 [15 pts]. Consider a p-n junction diode of Si as shown below:



(a) Find out the built in potential and the depletion width at $T=300\text{K}$.

[4 pt]

$$V_0 = V_T \ln \frac{N_A N_D}{n_i^2}$$

$$= 0.025 \ln \frac{10^{18} \times 10^{16}}{10^{20}}$$

$$= 0.025 \times 6.6$$

$$V_0 = 0.165 \text{ V}$$

$$W = \sqrt{\frac{2\epsilon_s}{q} \times \left(\frac{1}{N_A} + \frac{1}{N_D} \right) V_0}$$

$$= \sqrt{\frac{2 \times 10^{-12}}{1.6 \times 10^{-19}} \times \left(\frac{1}{10^{18}} + \frac{1}{10^{16}} \right) \times 0.165}$$

$$\approx 3 \times 10^{-5} \text{ cm}$$

(b) Assume that at $T=300\text{K}$ a voltage of 0.9 volt is applied across the diode such that the diode is forward biased. How much should the voltage have to be changed if at $T=400\text{K}$ one needs to ensure that the same current is flowing as in $T=300\text{K}$? Assume no change in bandgap due to change in temperature.

[5 pt]

Same current appears when:

$$V_1 - V_{01} = V_2 - V_{02}$$

$\underbrace{\hspace{2em}}_{\text{Built in pot}}$
 $\underbrace{\hspace{2em}}_{\text{Built in pot}}$

$$V_{02} = V_{T2} \ln \frac{N_A N_D}{n_i^2}$$

$$n_i = 5.2 \times 10^{15} \times (400)^{3/2} \times e^{-0.026 \times 400 / 0.025}$$

$$\approx 4 \times 10^{12}$$

$$T_2 = 400$$

$$V_{02} = \frac{0.025}{400} \times \ln \frac{10^{16} \times 10^{18}}{4 \times 10^{12}}$$

$$V_{02} = \frac{0.025}{400} \times 9.8$$

$$= 0.006125$$

$$V_2 = V_1 - V_{01} + V_{02}$$

$$= 0.9 - 0.165 + 0.006125$$

$$V_2 = 0.741 \text{ V}$$

- (c) In this problem we shall design a diode to meet certain requirements. We want a total current of 40 A/cm^2 at a voltage of 0.8 V at $T=300\text{K}$. Half of the current will have to be supplied by holes and half by electrons. Assume the mobility of electrons to be $500 \text{ cm}^2/\text{V-sec}$ and that of holes to be $250 \text{ cm}^2/\text{V-sec}$. Also assume that the diffusion length for electrons and holes are the same and equal to $10 \text{ }\mu\text{m}$. Find out how you will design your diode, i.e., how you will dope your p and n sides so that the aforementioned requirements are met. [6 pt]

$$J_n = \frac{e n_i^2}{L} \frac{D_n}{N_A} e^{-\frac{V_D}{V_T}} = \frac{40}{2}$$

$$\Rightarrow \frac{e n_i^2 V_T}{L} \frac{D_n}{N_A} e^{-\frac{V_D}{V_T}} = 20$$

$$N_A = \frac{1.6 \times 10^{19} \times 10^{-26} \times 26 \times 10^{-3}}{10 \times 10^{-4} \times 500 \times 10^{-13}} \times \frac{0.8}{0.026} \approx 26 \times 5 \times 10^{15}$$

see from P3.

$$N_A = 1.3 \times 10^{17} \text{ cm}^{-3}$$

$$J_p = \frac{e n_i^2}{L} \times \frac{\mu_p V_T}{N_D} e^{-\frac{V_D}{V_T}} = J_n$$

$$\frac{N_A}{N_D} = \frac{\mu_p \mu_n}{\mu_n}$$

$$N_D = \frac{N_A}{2}$$

Prob 3. [20 pts] Bipolar junction transistors.

(a) Design Fundamentals [4 pts].

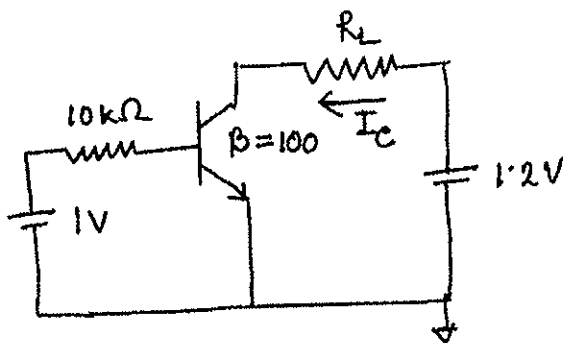
(i) Why is the base region doped more lightly than the emitter? [2 pt]

increases emitter to base flow
decreases base to emitter flow
hence increases gain.

(ii) Why is the width of the base region narrower than emitter and collector regions? [2 pt]

to minimize probability of recombination.

(b) Write down the condition for active mode operation and saturation mode operation for a NPN bipolar junction transistor. Consider the following transistor that has $\beta=100$. If it is desired that 1 mA current must be flowing in the collector at $V_{CC}=1$ V, what is the maximum value for R_L that can be used before the transistor goes from active to saturation mode? [5pt]



active: $V_C > V_E$

$V_C > V_B$

sat: $V_B > V_E$

$V_B > V_C$

Edge of sat:

$$V_B = V_C$$

$$V_B = 1V - I_B \times 10k\Omega$$

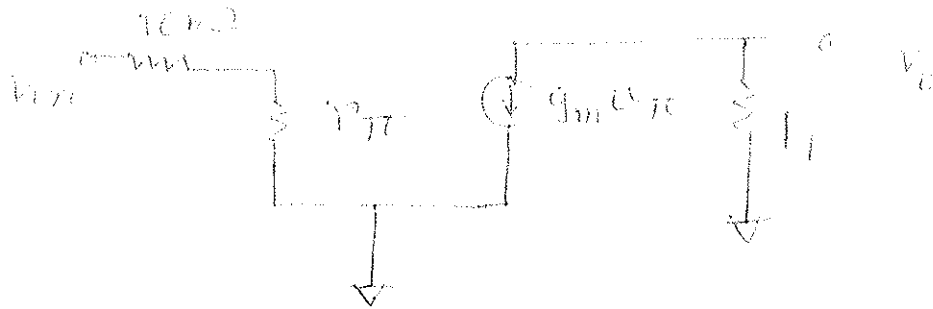
$$V_C = 1.2V - I_C \times R_L$$

$$1.2 - 10^{-3} R_L = 1 - \frac{10^{-3}}{100} \times 10 \times 10^3$$

$$\Rightarrow 10^{-3} R_L = 1.2 - 0.9$$

$$\Rightarrow \boxed{R_L = 300 \Omega} \quad \text{hence } \boxed{R_L \leq 300 \Omega}$$

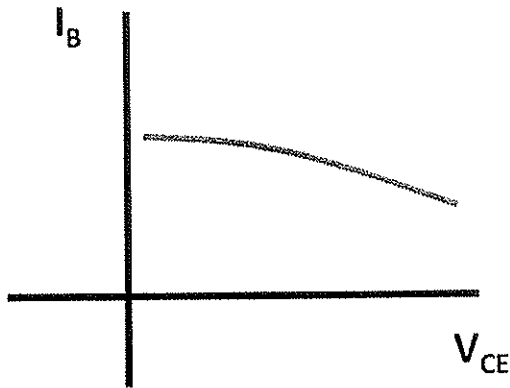
- (c) Draw the small signal model for the BJT described in part (b) assuming no Early effect. Indicate the numerical values and units for all the components of this small signal model. [5 pt]



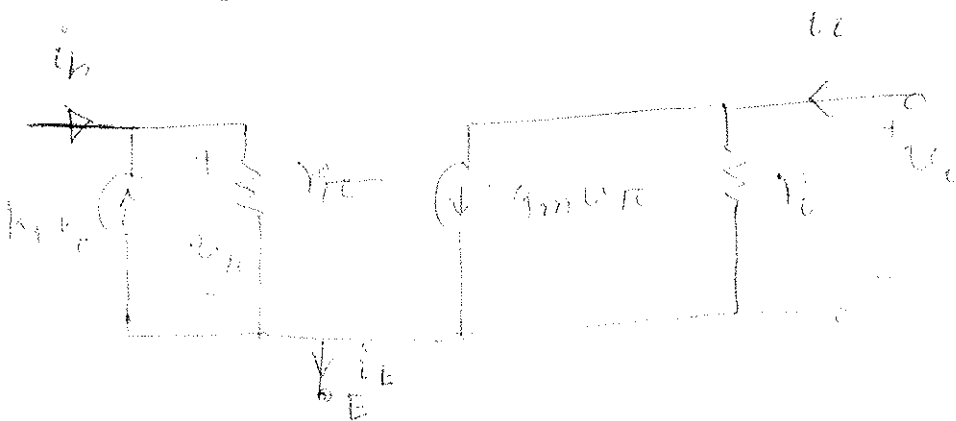
$$g_m = \frac{I_C}{V_T} = \frac{10^{-3}}{26 \times 10^{-3}} = \frac{1}{26} \Omega^{-1}$$

$$r_{\pi} = \frac{1}{g_m} = \frac{100}{1/26} = 2600 \Omega$$

- (d) Early effect/ base width modulation causes the collector current to increase with collector-emitter voltage, V_{CE} . Normally it is assumed that the base current remains independent of V_{CE} . However, if the depletion region spreads a long distance into the base, it can reduce the base current by recombination of injected holes with the negative immobile charges in the base region (for a NPN transistor). This is shown in the following figure. Considering this decrease of I_B with V_{CE} , derive and draw complete (including conventional early effect and the new phenomenon) the small-signal model of a NPN transistor. [6 pt]



Since $\frac{\partial I_B}{\partial V_{CE}} \neq 0$, we need to add a controlled source in the input circuit



$$k_1 = \frac{\partial I_B}{\partial V_{CE}}$$